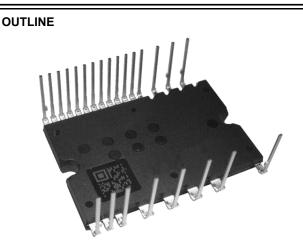


< DIPIPM > PSF25S92F6-A6

TRANSFER MOLDING TYPE INSULATED TYPE



MAIN FUNCTION AND RATINGS

- 3 phase DC/AC inverter
- 600V / 25A (MOSFET)
- N-side MOSFET open source
- Built-in bootstrap diodes with current limiting resistor

APPLICATION

 AC 100~240Vrms(DC voltage:400V or below) class low power motor control

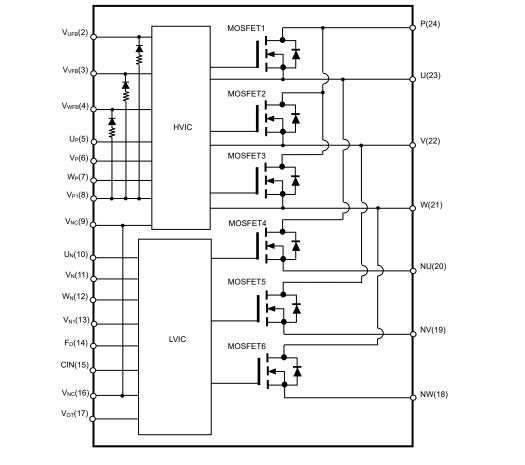
TYPE NAME

PSF25S92F6-A6	With temperature output function
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INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection
- For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC)
- Fault signaling : Corresponding to SC fault (N-side MOSFET), UV fault (N-side supply)
- Temperature output : Outputting LVIC temperature by analog signal
- Input interface : 3, 5V line, Schmitt trigger receiver circuit (High Active)
- UL Recognized : UL1557 File E323585

INTERNAL CIRCUIT



MAXIMUM RATINGS (T_{ch} = 25°C, unless otherwise noted)

INVERTER PART

Symbol	Parameter	Condition	Ratings	Unit
V _{DD}	Supply voltage	Applied between P-NU,NV,NW	450	V
V _{DD(surge)}	Supply voltage (surge)	Applied between P-NU,NV,NW	500	V
V _{DSS}	Drain-source voltage		600	V
±l _D	Each MOSFET drain current	T _C = 25°C (Note 1)	25	А
±l _{DP}	Each MOSFET drain current (peak)	T _C = 25°C, less than 1ms	50	А
T _{ch}	Channel temperature	(Note 2)	-30~+150	°C
Nata 1, Dula	a width and nariad are limited due to channel.	to man a rati ura		

Note1: Pulse width and period are limited due to channel temperature.

Note: The maximum channel temperature rating of built-in power chips is 150°C(@Tc≤100°C).However, to ensure safe operation of DIPIPM, the average channel temperature should be limited to Tch(Ave)≤125°C (@Tc≤100°C).

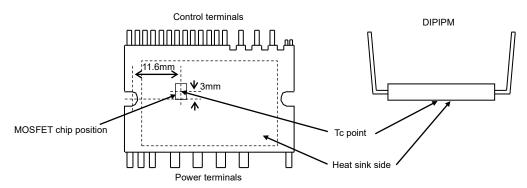
CONTROL (PROTECTION) PART

Symbol	Parameter	Condition	Ratings	Unit
VD	Control supply voltage	Applied between V _{P1} -V _{NC} , V _{N1} -V _{NC}	24	V
V _{DB}	Control supply voltage	Applied between V _{UFB} -U, V _{VFB} -V, V _{WFB} -W	24	V
VIN	Input voltage	Applied between U _P , V _P , W _P , U _N , V _N , W _N -V _{NC}	-0.5~V _D +0.5	V
V _{FO}	Fault output supply voltage	Applied between Fo-VNC	-0.5~V _D +0.5	V
I _{FO}	Fault output current	Sink current at F _o terminal	1	mA
Vsc	Current sensing input voltage	Applied between CIN-V _{NC}	-0.5~V _D +0.5	V

TOTAL SYSTEM

Symbol	Parameter Condition		Ratings	Unit
V _{DD(PROT)}	Self protection supply voltage limit $V_D = 17 \sim 19V$, Inverter Part(Short circuit protection capability) T_{ch} = 125°C, non-repetitive, less than 2µs		400	V
Tc	Module case operation temperature Measurement point of Tc is provided in Fig.1		-30~+100	°C
T _{stg}	Storage temperature		-40~+125	°C
V _{iso}	Isolation voltage	60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate	1500	V _{rms}

Fig. 1: T_c MEASUREMENT POINT



THERMAL RESISTANCE

Symbol Parameter		Condition		Limits			
				Тур.	Max.	Unit	
R _{th(ch-c)Q}	Channel to case thermal resistance	Inverter MOSFET part (per 1/6 module)	-	-	2.0	K/W	

Note 3: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100µm~+200µm on the contacting surface of DIPIPM and heat sink. The contacting thermal resistance between DIPIPM case and heat sink Rth(c-f) is determined by the thickness and the thermal conductivity of the applied grease. For reference, Rth(c-f) is about 0.3K/W (per 1/6 module, grease thickness: 20µm, thermal conductivity: 1.0W/m·K).

ELECTRICAL CHARACTERISTICS (T_{ch} = 25°C, unless otherwise noted) **INVERTER PART**

Currada al	Deveryoter	Cand	141 a.m.		Limits			Unit
Symbol	Parameter	Cond	llion		Min.	Тур.	Max.	Unit
V	Drain-source on-state	(1 - 1) = 10 (1) = 5 (1)	I _D = 25/	A, T _{ch} = 25°C	-	1.30	2.20	v
V _{DS(on)}	voltage	$V_{\rm D} = V_{\rm DB} = 18V, V_{\rm IN} = 5V$	I _D = 25/	A, T _{ch} = 125°C	-	1.20	2.00	v
$V_{\text{SD(off)}}$	Source- drain voltage	$V_{D}=V_{DB} = 18V, V_{IN} = 0V, -I_{D} = 25A$		-	4.00	5.00	V	
t _{on}		V _{DD} = 300V, V _D = V _{DB} = 18V			1.00	1.50	1.95	μs
t _{c(on)}				-	0.25	0.50	μs	
t _{off}	Switching times	imes $I_D = 25A, T_{ch} = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ Inductive Load (upper-lower arm)	-	1.70	2.20	μs		
t _{c(off)}				-	0.15	0.45	μs	
t _{rr}	1			-	0.05	-	μs	
	Drain-Source			$T_{ch} = 25^{\circ}C$	-	-	1	m۸
IDSS	cut-off current	V _{DS} =V _{DSS}		T _{ch} =125°C	-	-	10	mA

CONTROL (PROTECTION) PART

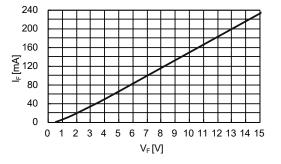
Sumbol	Parameter	Cond	ition		Unit		
Symbol	Parameter	Cond	luon	Min.	Тур.	Max.	Unit
			V _D =18V, V _{IN} =0V	-	-	3.50	
I _D	- Circuit current	Total of V_{P1} - V_{NC} , V_{N1} - V_{NC}	V _D =18V, V _{IN} =5V	-	-	3.50	mA
		Each part of V _{UFB} -U,	V _D =V _{DB} =18V, V _{IN} =0V	-	-	0.35	IIIA
I _{DB}		V _{VFB} -V, V _{WFB} -W	$V_D = V_{DB} = 18V$, $V_{IN} = 5V$	-	-	0.35	
V _{SC(ref)}	Short circuit trip level	V _D = 18V	(Note 4)	0.455	0.480	0.505	V
UV _{DBt}	P-side Control supply		Trip level	10.0	-	12.0	V
UV_{DBr}	under-voltage protection(UV)	T _{ch} ≤125°C	Reset level	10.5	-	12.5	V
UV _{Dt}	N-side Control supply	T _{ch} = 120 C	Trip level	10.3	-	12.5	V
UV _{Dr}	under-voltage protection(UV)		Reset level	10.8	-	13.0	V
V	Temperature Output	Pull down R=5.1kΩ (Note 5)	LVIC Temperature=90°C	2.63	2.77	2.91	V
V _{OT}		Pull down R=5.1kΩ (Note 5) LVIC Temperature=25°C		0.88	1.13	1.39	V
V_{FOH}	- Fault output voltage	V_{SC} = 0V, F_{O} terminal pulled up	o to 5V by 10kΩ	4.9	-	-	V
V _{FOL}	Fault output voltage	V_{SC} = 1V, I_{FO} = 1mA		-	-	0.95	V
t _{FO}	Fault output pulse width		(Note 6)	20	-	-	μs
I _{IN}	Input current	$V_{IN} = 5V$		0.70	1.00	1.50	mA
V _{th(on)}	ON threshold voltage			-	2.10	2.60	
$V_{\text{th(off)}}$	OFF threshold voltage	Applied between U _P , V _P , W _P , U _N , V _N , W _N -V _{NC}		0.80	1.50	-	v
$V_{\text{th(hys)}}$	ON/OFF threshold hysteresis voltage			0.35	0.65	-	
V _F	Bootstrap Di forward voltage	IF=10mA including voltage drop b	by limiting resistor (Note 7)	0.9	1.3	1.7	V
R	Built-in limiting resistance	Included in bootstrap Di		48	60	72	Ω

Note 4 : SC protection works only for N-side MOSFETs. Please select the external shunt resistance such that the SC trip-level is less than 1.7 times of the current rating.

5 : For temperature output type, DIPIPM don't shutdown MOSFETs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that user defined, controller (MCU) should stop the DIPIPM. Temperature of LVIC vs. VOT output characteristics is described in Fig. 3.

6 : Fault signal Fo outputs when SC or UV protection works. Fo pulse width is different for each protection modes. At SC failure, Fo pulse width is a fixed width (=minimum 20μs), but at UV failure, Fo outputs continuously until recovering from UV state. (But minimum Fo pulse width is 20μs.)
 7 : The characteristics of bootstrap Di is described in Fig.2.

Fig. 2 Characteristics of bootstrap Di V_F-I_F curve (@Ta=25°C) including voltage drop by limiting resistor (Right chart is enlarged chart.)



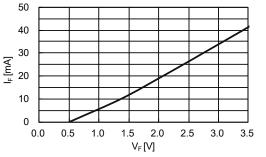


Fig. 3 Temperature of LVIC vs. V_{OT} output characteristics

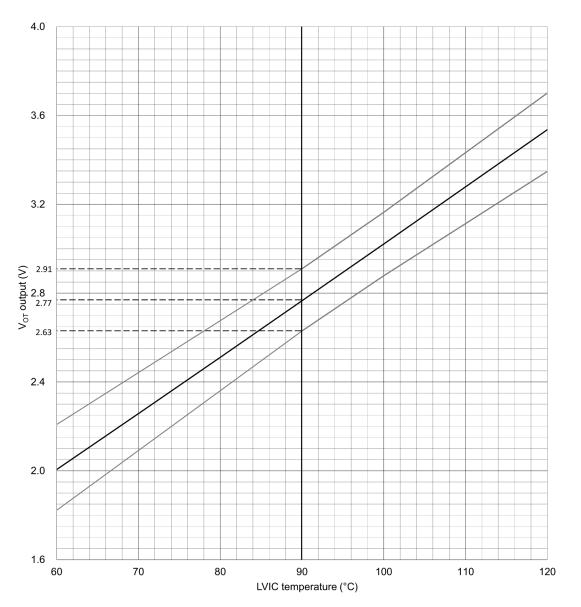
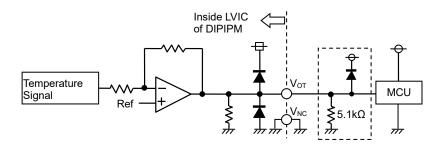


Fig. 4 Vor output circuit



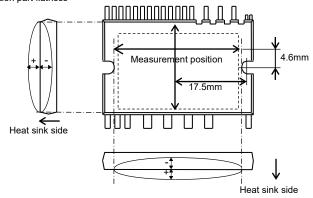
- (1) It is recommended to insert $5k\Omega$ (5.1 $k\Omega$ is recommended) pull down resistor for getting linear output characteristics at low temperature below room temperature. When the pull down resistor is inserted between V_{OT} and V_{NC}(control GND), the extra circuit current, which is calculated approximately by V_{OT} output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using V_{OT} for detecting high temperature over room temperature only, it is unnecessary to insert the pull down resistor.
- (2) In the case of using V_{OT} with low voltage controller like 3.3V MCU, V_{OT} output might exceed control supply voltage 3.3V when temperature rises excessively. If system uses low voltage controller, it is recommended to insert a clamp Di between control supply of the controller and V_{OT} output for preventing over voltage destruction.
- (3) In the case of not using $V_{\text{OT}},$ leave V_{OT} output NC (No Connection).

Refer the application note about the usage of $V_{\mbox{\scriptsize OT}}.$

MECHANICAL CHARACTERISTICS AND RATINGS

Parameter	Condition	Reference		Unit		
Falameter	Condition	Relefence	Min.	Тур.	Max.	Unit
Mounting torque	Mounting screw : M3 (Note 8)	JEITA-ED-4701 402 method II	0.59	0.69	0.78	N∙m
Terminal pulling strength	Control terminal: Load 5N Power terminal: Load 10N	JEITA-ED-4701 401 method I	10	-	-	s
Terminal bending strength	Control terminal: Load 2.5N Power terminal: Load 5N 90deg. bend	JEITA-ED-4701 401 method III	2	-	-	times
Weight			-	8.5	-	g
Heat radiation part flatness	(Note 9)		-50	-	100	μm

Note 8: Plain washers (ISO 7089~7094) are recommended. Note 9: Measurement point of heat radiation part flatness



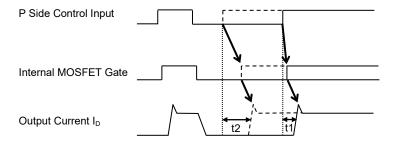
RECOMMENDED OPERATION CONDITIONS

Symbol	Parameter	Con	dition		Limits		
Symbol	Parameter	Condition		Min.	Тур.	Max.	Unit
V _{DD}	Supply voltage	Applied between P-NU, NV	γ, NW	0	300	400	V
VD	Control supply voltage	Applied between V _{P1} -V _{NC} , V	/ _{N1} -V _{NC}	17.0	18.0	19.0	V
V _{DB}	Control supply voltage	Applied between V _{UFB} -U, V	_{VFB} -V, V _{WFB} -W	15.0	18.0	22.0	V
$\Delta V_D, \Delta V_{DB}$	Control supply variation			-1	-	+1	V/µs
t _{dead}	Arm shoot-through blocking time	For each input signal		1.5	-	-	μs
f _{PWM}	PWM input frequency	T _C ≤ 100°C, T _{ch} ≤ 125°C		-	5	20	kHz
PWIN(on)		(Note 10)		0.7	-	-	
PWIN(off)	Minimum input pulse width	$\begin{array}{l} 200V{\leq}V_{\text{DD}}{\leq}350V,\\ 17V{\leq}V_{\text{D}}{\leq}19V,\\ 15V{\leq}V_{\text{DB}}{\leq}22V,\\ -30^{\circ}C{\leq}\text{Tc}{\leq}100^{\circ}\text{C},\\ \text{N-line wiring inductance}\\ \text{less than 10nH (Note 11)} \end{array}$	Between rated current and 1.7 times of rated current	1.5	-	-	μs
V _{NC}	V _{NC} variation	Between V _{NC} -NU, NV, NW (including surge)		-5	-	+5	V
T_{ch}	Channel temperature			-30	-	+125	°C

Note 10: DIPIPM might not make response if the input signal pulse width is less than PWIN(on).

11: IPM might make delayed response or no response for the input signal with off pulse width less than PWIN(off). Please refer below about delayed response.

Delayed Response against Shorter Input Off Signal than PWIN(off) (P-side only)

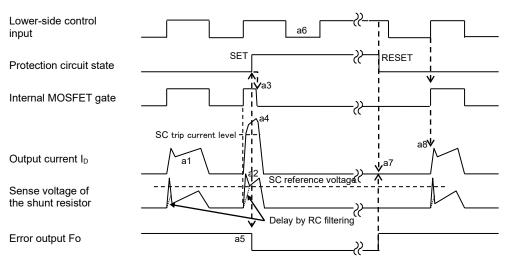


Real line: off pulse width > PWIN(off); turn on time t1 Broken line: off pulse width < PWIN(off); turn on time t2 (t1:Normal switching time)

Fig. 5 Timing Charts of The DIPIPM Protective Functions

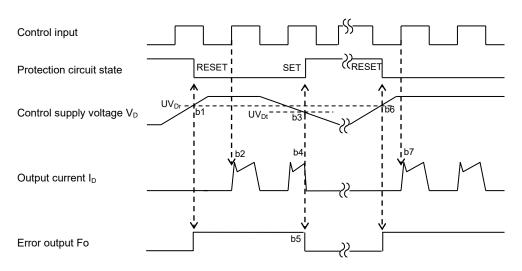
[A] Short-Circuit Protection (N-side only with the external shunt resistor and RC filter)

- a1. Normal operation: MOSFET ON and outputs current.
- a2. Short circuit current detection (SC trigger)
 - (It is recommended to set RC time constant 1.5~2.0µs so that MOSFET shut down within 2.0µs when SC.)
- a3. All N-side MOSFET's gates are hard interrupted.
- a4. All N-side MOSFETs turn OFF.
- a5. F_{O} outputs for $t_{\text{Fo}}\text{=}minimum$ 20µs.
- a6. Input = "L": MOSFET OFF
- a7. Fo finishes output, but MOSFETs don't turn on until inputting next ON signal (L→H). (MOSFET of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: MOSFET ON and outputs current.



[B] Under-Voltage Protection (N-side, UV_D)

- b1. Control supply voltage V_D exceeds under voltage reset level (UV_{Dr}), but MOSFET turns ON by next ON signal (L→H). (MOSFET of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: MOSFET ON and outputs current.
- b3. V_D level drops to under voltage trip level. (UV_{Dt}).
- b4. All N-side MOSFETs turn OFF in spite of control input condition.
- b5. Fo outputs for t_{Fo} =minimum 20µs, but output is extended during V_D keeps below UV_{Dr}.
- b6. V_D level reaches UV_{Dr} .
- b7. Normal operation: MOSFET ON and outputs current.



[C] Under-Voltage Protection (P-side, UV_{DB})

- c1. Control supply voltage V_{DB} rises. After the voltage reaches under voltage reset level UV_{DBr}, MOSFET turns on by next ON signal (L \rightarrow H).
- c2. Normal operation: MOSFET ON and outputs current.
- c3. V_{DB} level drops to under voltage trip level (UV_{DBt}).
- c4. MOSFET of the corresponding phase only turns OFF in spite of control input signal level, but there is no F_0 signal output.
- c5. V_{DB} level reaches UV_{DBr} .
- c6. Normal operation: MOSFET ON and outputs current.

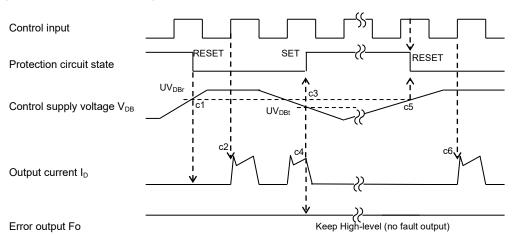
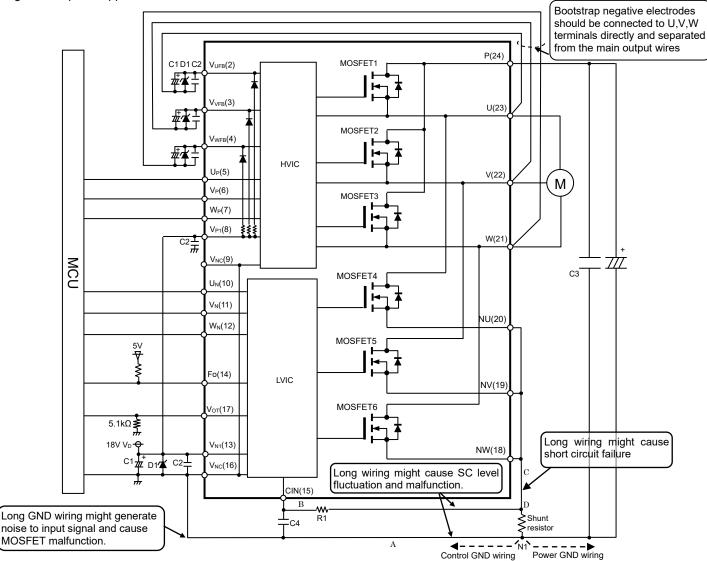


Fig. 6 Example of Application Circuit



- (1) If control GND is connected with power GND by common broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect control GND and power GND at only a point N1 (near the terminal of shunt resistor).
- (2) It is recommended to insert a Zener diode D1(24V/1W) between each pair of control supply terminals to prevent surge destruction.
 (3) To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Generally a 0.1-0.22µF snubber capacitor C3 between the P-N1 terminals is recommended.
- (4) R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2μs. (1.5μs~2μs is general value.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the rea 1 5 VI system is necessary.
- (5) To prevent malfunction, the wiring of A, B, C should be as short as possible.
- (6) The point D at which the wiring to CIN filter is divided should be near the terminal of shunt resistor. NU, NV, NW terminals should be connected at near NU, NV, NW terminals.
- (7) All capacitors should be mounted as close to the terminals as possible. (C1: good temperature, frequency characteristic electrolytic type and C2:0.22µ-2µF, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- (8) Input drive is High-active type. There is a minimum 3.3kΩ pull-down resistor in the input circuit of IC. To prevent malfunction, the wiring of each input should be as short as possible. When using RC coupling circuit, make sure the input signal level meet the turn-on and turn-off threshold voltage.
- (9) Fo output is open drain type. It should be pulled up to MCU or control power supply (e.g. 5V,18V) by a resistor that makes I_{Fo} up to 1mA. (I_{FO} is estimated roughly by the formula of control power supply voltage divided by pull-up resistance. In the case of pulled up to 5V, 10kΩ (5.1kΩ or more) is recommended.)
- (10) Thanks to built-in HVIC, direct coupling to MCU without any opto-coupler or transformer isolation is possible.
- (11) Two V_{NC} terminals (9 & 16 pin) are connected inside DIPIPM, please connect either one to the 18V power supply GND outside and leave another one open.
- (12) If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause DIPIPM erroneous operation. To avoid such problem, line ripple voltage should meet dV/dt ≤+/-1V/µs, Vripple≤2Vp-p.
- (13) For DIPIPM, it isn't recommended to drive same load by parallel connection with other phase MOSFET or other DIPIPM.

Fig. 7 MCU I/O Interface Circuit

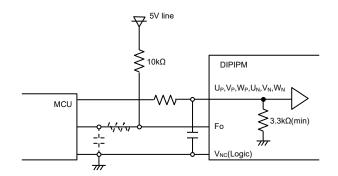
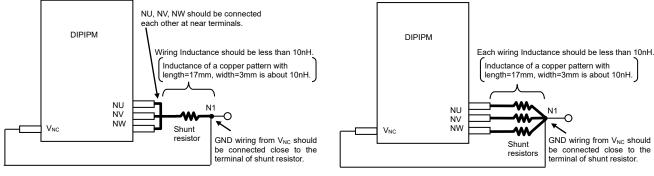


Fig. 8 Pattern Wiring Around the Shunt Resistor

Note)

Design for input RC filter depends on PWM control scheme used in the application and wiring impedance of the printed circuit board. DIPIPM input signal interface integrates a minimum $3.3 k\Omega$ pull-down resistor. Therefore, when inserting RC filter, it is necessary to satisfy turn-on threshold voltage requirement.

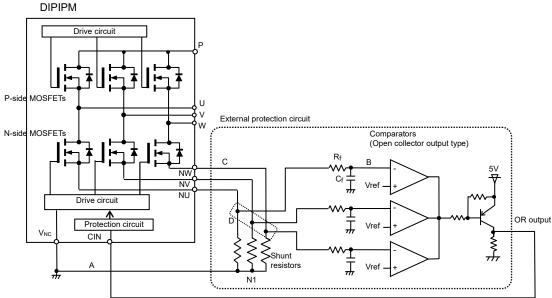
Fo output is open drain type. It should be pulled up to the positive side of 5V or 18V power supply with the resistor that limits Fo sink current I_{Fo} under 1mA. In the case of pulling up to 5V supply, over 5.1k Ω is needed. (10k Ω is recommended.)



Low inductance shunt resistor like surface mounted (SMD) type is recommended.

Fig. 9 Pattern Wiring Around the Shunt Resistor (for the case of open emitter)

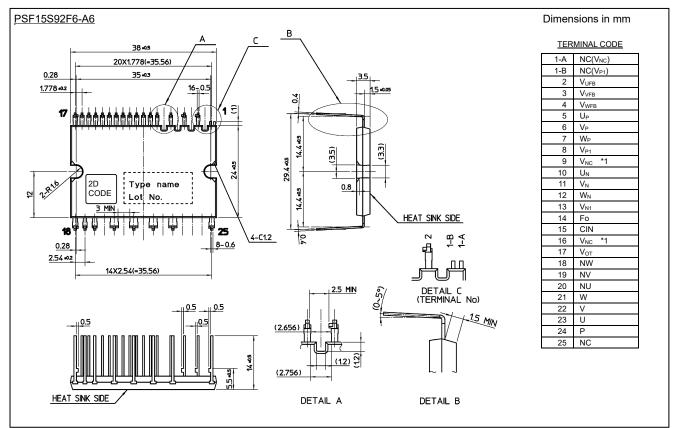
When DIPIPM is operated with three shunt resistors, voltage of each shunt resistor cannot be input to CIN terminal directly. In that case, it is necessary to use the external protection circuit as below.



(1) It is necessary to set the time constant R_fC_f of external comparator input so that MOSFET stops within 2µs when short circuit occurs. SC interrupting time might vary with the wiring pattern, comparator speed and so on.

- (2) It is recommended for the threshold voltage Vref to set to the same rating of short circuit trip level (Vsc(ref): typ. 0.48V).
- (3) Select the external shunt resistance so that SC trip-level is less than specified value (=1.7 times of rating current).
- (4) To avoid malfunction, the wiring A, B, C should be as short as possible.
- (5) The point D at which the wiring to comparator is divided should be close to the terminal of shunt resistor.
- (6) OR output high level when protection works should be over 0.505V (=maximum Vsc(ref) rating).

Fig. 10 Package Outlines



1) 9 & 16 pins (V_{NC}) are connected inside DIPIPM, please connect either one to the control power supply GND outside and leave another one open.

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